# A Communication-Bandwidth-Aware Hybrid Estimation Framework for Multi-robot Cooperative Localization

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Abstract—This paper presents hybrid Minimum Mean Squared Error estimators for wireless sensor networks with time-varying communication-bandwidth constraints, focusing on the particular application of multi-robot Cooperative Localization. When sensor nodes (e.g., robots) communicate only a quantized version of their analog measurements to the team, our proposed hybrid filters enable robots to process all available information, i.e., local analog measurements (recorded by its own sensors) as well as remote quantized measurements (collected and communicated by other sensors). Moreover, these filters are resource-aware and can utilize additional bandwidth, whenever available, to maximize estimation accuracy. Specifically, in this paper, we present two filters, the Hybrid Batch-Quantized Kalman filter (H-BQKF) and the Hybrid Iteratively-Quantized Kalman filter (H-IQKF), that can process local analog measurements along with remote measurements quantized to any number of bits. We test our proposed filters in simulations and experimentally, and demonstrate that they achieve performance comparable to the standard Kalman filter.

# I. INTRODUCTION AND RELATED WORK

In wireless sensor (robot) network (WSN) applications (e.g., target tracking and environmental monitoring), sensors typically estimate a quantity of interest, using noisy measurements from *all* sensor nodes. Since the sensor nodes are spatially distributed, each node generally has to communicate its local information to the team, either as estimate-covariance pairs or as raw measurements [1], incurring substantial communication overhead. Therefore, for WSNs deployed in (i) environments with inherent communication limitations or (ii) applications with power/battery restrictions, it becomes necessary to develop decentralized estimation algorithms that can trade communication bandwidth-availability for estimation accuracy. While bandwidth constraints exist in many WSN applications, this work focuses on the representative application of Cooperative Localization (CL).

In GPS-denied environments (e.g., space or indoors), CL is used for accurate multi-robot localization, i.e., precisely estimating the robots' poses (position and orientation). In CL, communicating robots equipped with proprioceptive (e.g., wheel encoders) and exteroceptive (e.g., cameras) sensors use their individual motion measurements (e.g., linear/rotational velocity) and robot-to-robot relative measurements (e.g., distance/bearing) to *jointly* estimate their poses, resulting in increased accuracy for the entire team [2].

Representative works for bandwith-constrained CL include Extended Kalman Filter (EKF)-based approaches of [3], [4],

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and [5], where based on a suitable selection/optimality criterion, the robots select and transmit their most informative analog measurements<sup>1</sup>. Similarly, Maximum A Posterior (MAP) estimator-based approaches for CL, where the robots only periodically exchange local information, are presented in [6], [7]. However, in contrast to these approaches, which assume that robots can communicate all or a subset of their analog measurements, the work presented in this paper focuses on applications with stringent communication bandwidth constraints, where robots can communicate only a few bits per analog measurement. Therefore, each robot has to perform lossy quantization of its analog measurements before communicating them to the team. Moreover, the existing estimators such as the EKF and MAP, designed for processing only analog measurements, have to be modified to handle the quantized measurements.

Estimation with quantized observations has been wellstudied in the signal processing community for WSNs. While there exists a large body of work on parameter estimation (either deterministic [8], [9], [10] or random variable [11], [12], [13]), we will focus on approaches that were developed to estimate random processes, as is the case in CL. The Sign-of-Innovation Kalman filter (SOI-KF) [14], has been proposed for estimating stochastic dynamic processes, where the measurement innovation<sup>2</sup>, instead of the actual analog measurement, is quantized to a single bit. Developed for linear and Gaussian process and measurement models, the SOI-KF approximates the posterior probability density function (pdf) by a Gaussian<sup>3</sup> after each measurement update, resulting in a recursive state/covariance update structure very similar to that of the standard Kalman filter (KF) [15]. When additional bits are available for quantization, the SOI-KF approach has been extended in [16] to the batch-quantized KF (BQKF) and the iteratively-quantized KF (IQKF), where performance comparable to the standard KF can be achieved by communicating only 4 bits per analog measurement. An extension to quantized batch MAP estimation, that improves estimation accuracy when using nonlinear system models, is presented in [17].

However, the main drawback of these quantized estimation schemes is that they prohibit the robots from utilizing all

<sup>&</sup>lt;sup>1</sup>Sensors sample a process and provide a measurement which is often represented in digital form using 32 or 64 bit floating-point number representation. We refer to such measurements as analog.

<sup>&</sup>lt;sup>2</sup>Measurement innovation is the difference between the actual and the estimated (by the filter) measurement.

<sup>&</sup>lt;sup>3</sup>Note that due to the nonlinearity of the quantization operation, the posterior pdf is not, in general, a Gaussian.

locally-available measurement information. Specifically, to ensure estimation consistency, each robot is compelled to use only quantized versions of its *own* local, analog measurements. Thus, valuable information, that can be used to improve localization accuracy, is discarded.

In [18], we introduced a hybrid estimation framework that addresses this problem by maintaining two local estimators for each robot (see Fig. 1): (i) a quantized Q-estimator processing quantized measurements from all robots, including itself, and (ii) a hybrid H-estimator processing its own analog measurements along with the quantized measurements from other robots in the team. The H-estimator in [18] is designed for the special/restrictive scenario when robots communicate *only a single bit* per analog measurement.

However, in practice, the communication bandwidth available for CL is often time-varying, and depends upon the resource requirements of higher-level tasks and the robots' battery life. Therefore, in this work, our objective is to efficiently utilize additional bandwidth, whenever available, to maximize localization accuracy. To achieve this, in this paper, we derive H-estimators that can handle the general case of time-varying communication-bandwidth availability, i.e., when robots in the team can communicate  $f \geq 1$ bits, per analog measurement. Specifically, we develop Hestimators (see Section III) for two quantization scenarios: (i) Batch quantization: where the bandwidth availability is known beforehand, and (ii) Iterative quantization: where additional bandwidth becomes available on-the-fly. For both these scenarios, we derive Minimum Mean Squared Error (MMSE) H-estimators (H-BQKF and H-IQKF, respectively) that are capable of processing local analog measurements along with multiple bits per remote analog measurement. Lastly, in Section IV, we present extensive simulation and experimental results that study the performance and accuracy of the proposed H-estimators.

# II. PROBLEM FORMULATION

The proposed hybrid MMSE estimators are designed for wireless sensor networks where (i) the process and measurement models are shared *a priori* by all sensor nodes, and (ii) each sensor node can communicate with the network at every time step. While we now proceed with the specific application of CL, we note that the proposed estimators are general and can be used for any static/mobile sensor network applications that satisfy the above assumptions.

For CL, the problem setup consists of a team of N robots performing multi-centralized CL (MC-CL) in 2D. In MC-CL, each robot broadcasts all its measurements and every robot locally processes measurements from the entire team to generate pose estimates for all the robots. The robot team uses a statistical motion model (e.g., constant-velocity model [19]), driven by system noise, as the process model:

$$\mathbf{x}_k = \mathbf{F}_{k-1} \mathbf{x}_{k-1} + \mathbf{G}_{k-1} \mathbf{w}_{k-1}, \ \mathbf{x}_0 \sim \mathcal{N}(\mathbf{x}_{init}, \mathbf{P_0}) \quad (1)$$

where,  $\mathbf{w}_k$  is the zero-mean, white, Gaussian, and uncorrelated system noise at time-step k with covariance  $E\left[\mathbf{w}_k\mathbf{w}_l^T\right] = \delta_{kl}\mathbf{Q}_k$ . Here,  $\mathbf{x}_k = [\mathbf{x}_k^{1T}, \mathbf{x}_k^{2T}, \dots, \mathbf{x}_k^{NT}]^T$ ,

is the joint-state of the team and  $\mathbf{x}_k^i = [x_k^i, y_k^i, \phi_k^i]^T$ , is the state (position and orientation) of robot i at time-step k.

Robot i obtains  $M_k^i$  scalar, analog measurements (proprioceptive and exteroceptive) at time-step k. The measurement model for robot i, i = 1, ..., N, is:

$$z_{km}^{i} = \mathbf{h}_{km}^{i^{T}} \mathbf{x}_{k} + v_{km}^{i}, \ m = 1, \dots, M_{k}^{i}$$
 (2)

where  $v^i_{km}$  is zero-mean, white, Gaussian, and uncorrelated measurement noise with  $E[v^i_{km}v^i_{ln}]=\delta_{kl,mn}\sigma^{i^2}_{km}$  and  $E[v^i_{km}v^j_{ln}]=0, \ \forall j\neq i, \ j=1,\ldots,N.$  Here, the linear models (1)-(2) are used only for mathematical derivations. In real-world scenarios with nonlinear models (see Section IV), the corresponding linearized system models are used. Also, to simplify the notation in the paper, we will assume that each robot i obtains only a single, scalar, analog measurement,  $z^i_k$ , at time-step k. The generalization to  $M^i_k$  measurements is straightforward.

In the absence of communication-bandwidth constraints, the robots broadcast their analog measurements to the team. Then each robot uses the standard KF [15] to obtain the Minimum Mean Squared Error (MMSE) estimates of all robots' poses at time-step k, given all analog measurements,  $\mathbf{z}_{0:k}$ , up to time-step k. Here,  $\mathbf{z}_{0:k} = [(\mathbf{z}_{0:k}^1)^T, \ldots, (\mathbf{z}_{0:k}^N)^T]^T$  and  $\mathbf{z}_{0:k}^i = [z_0^i, \ldots, z_k^i]^T$ ,  $i = 1, \ldots, N$ . On the contrary, in the bandwidth-limited scenario, each robot can communicate only  $f \geq 1$  bits per analog measurement. Therefore, robot i quantizes its analog measurement,  $z_k^i \in \mathbb{R}$ , to  $b_k^i \in \mathcal{B}$ ,  $\mathcal{B} := \{1, \ldots, 2^f\}$  using the following quantization rule before broadcasting it:

$$b_k^i = \mathbf{q}[z_k^i], \text{ where } \mathbf{q} : \mathbb{R} \to \mathcal{B}.$$
 (3)

Next, each robot uses a quantized filter, such as the BQKF or the IQKF [16], to generate MMSE estimates (under Gaussian assumption<sup>4</sup>) for the robots' poses using all quantized measurements,  $\mathbf{b}_{0:k}$ , up to time-step k. However, the quantization rules used by these filters depend upon the robots' pose estimates, via the measurement innovation, and thus all robots have to maintain *identical* filters to ensure estimation consistency. Therefore, even though each robot i has access to its own analog measurements,  $\mathbf{z}_{0:k}^i$ , it is forced to discard this information and process only the corresponding quantized measurements,  $\mathbf{b}_{0:k}^i$ .

# III. HYBRID ESTIMATION FRAMEWORK

In order to address this problem, we introduced a novel hybrid estimation framework in [18] that enables each robot to process all available measurement information, i.e., local analog and remote quantized measurements. To achieve this, each robot i maintains two estimators (see Fig. 1): (1) a quantized Q-estimator that processes quantized measurements from all robots including itself, i.e.,  $\hat{\mathbf{x}}_{k|k}^Q := E[\mathbf{x}_k|\mathbf{b}_{0:k}^{q\neq i},\mathbf{b}_{0:k}^i]$ , and (2) a hybrid H-estimator that processes

<sup>4</sup>It is important to note that this loss of Gaussianity in the IQKF and BQKF is due to the non-linearity of the quantization step as opposed to the non-linearity of the process and measurement models in the EKF.

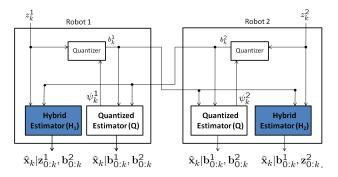


Fig. 1. Hybrid Estimation framework. When  $\psi_k^i = E[z_k^i|\mathbf{b}_{0:k-1}]$  and we use the quantization rule in (4), the H- and Q-estimators correspond to the H-BQKF and BQKF, respectively. When  $\psi_k^i = E[z_k^i|\mathbf{b}_{0:k-1},\mathbf{b}_k^{i(1:p-1)}]$  and we use the quantization rule in (7), the H- and Q-estimators correspond to the H-IQKF and IQKF, respectively, i=1,2.

its own real-valued measurements and quantized measurements from the other robots in the team, i.e.,  $\hat{\mathbf{x}}_{k|k}^{H_i} := E[\mathbf{x}_k|\mathbf{b}_{0:k}^{q\neq i},\mathbf{z}_{0:k}^i],\ q=1,\ldots,N.$  The estimates generated by the Q-estimator are *identical* for all robots since each robot processes the same measurements,  $\mathbf{b}_{0:k} = \left\{\mathbf{b}_{0:k}^{q\neq i},\mathbf{b}_{0:k}^i\right\}$ . Therefore, they are used in the hybrid estimation framework for generating quantization thresholds for all robots as shown in the next section. On the contrary, the estimates generated by the H-estimator are *different* for each robot, since each robot processes a different set of quantized and analog measurements.

Note that [18] considered the special, restrictive case where the robots communicate only one bit per analog measurement. In the next section, we derive MMSE (under Gaussian assumption) H-estimators that process: (i) local analog measurements, and (ii) multiple bits  $(f \geq 1)$ , per analog measurement, communicated by other robots in the team. Specifically, we derive H-estimators for two quantization scenarios: (1) the H-BQKF for batch quantization, where the bandwidth availability (f bits per analog measurement) is known a priori, and (2) H-IQKF for iterative quantization, when additional bandwidth becomes available on-the-fly.

### A. Batch Quantization

1) Encoding rule (Quantizer design): Since robot j is preinformed about the availability of  $f \geq 1$  bits for communicating its analog measurement  $z_k^j \in \mathbb{R}$ , robot j partitions the observation space  $\mathbb{R}$  into  $2^f$  intervals. The interval  $\mathcal{R}_k^j(n) := [\tau_k^j(n), \tau_k^j(n+1))$ , where  $\tau_k^j(n)$  are the quantization thresholds,  $n \in \mathcal{B} := \{1, \dots, 2^f\}, \, \tau_k^j(1) = -\infty, \, \tau_k^j(2^f+1) = \infty$ , and  $\tau_k^j(n) < \tau_k^j(n+1)$ . The quantization rule, which is based on the measurement innovation, has the form<sup>5</sup> [16]:

$$b_k^j = n$$
, iff  $(z_k^j - E[z_k^j | \mathbf{b}_{0:k-1}, \mathbf{b}_k^m]) \in [\tau_k^j(n), \tau_k^j(n+1))$  (4)

where,  $\mathbf{b}_{0:k-1}$  denotes the quantized measurements from all robots up to time-step k-1, and  $\mathbf{b}_k^m$  denotes the

quantized measurements from robot  $m, m = 1, \ldots, (j-1)$ , at time-step k. From (2), the predicted measurement is  $E[z_k^j|\mathbf{b}_{0:k-1},\mathbf{b}_k^m] =: \mathbf{h}_k^{j^T}\hat{\mathbf{x}}_{k|k,j-1}^Q$ . Note that robot j uses the Q-estimator's predicted measurement, that is identical for all robots, for generating the quantized measurements. This allows other robots to correctly process/decode the quantized measurement. Robot j does not use the predicted measurement of its H-estimator,  $E[z_k^j|\mathbf{b}_{0:k-1}^{q\neq j},\mathbf{b}_k^m,\mathbf{z}_{0:k}^j], q=1,\ldots,N$ , since it depends upon its local analog measurements,  $\mathbf{z}_{0:k}^j$ , which are unavailable to the other robots.

2) Decoding rule (Estimator design): For the batch quantization rule from (4), we now derive the resulting MMSE Q-and H-estimators. Note that the Q-estimator, by definition, is identical to the BQKF in [16] and is not presented here due to space constraints. The H-estimator, H-BQKF, is obtained as follows.

Proposition 1: H-estimator (H-BQKF)

Consider the linear model of (1)-(2) and the quantization rule in (4). If robot i assumes the pdf  $p\left[\mathbf{x}_{k-1}|\mathbf{b}_{0:k-1}^{q\neq i},\mathbf{z}_{0:k-1}^{i}\right]\sim\mathcal{N}\left(\hat{\mathbf{x}}_{k-1|k-1}^{H_{i}},\mathbf{P}_{k-1|k-1}^{H_{i}}\right)$ , then the state/covariance propagation equations from time-step k to k+1 are identical to the KF. If robot i, assumes the pdf  $p\left[\mathbf{x}_{k}|\mathbf{b}_{0:k-1}^{q\neq i},\mathbf{b}_{k}^{m\neq i},\mathbf{z}_{0:k-1}^{i}\right]\sim\mathcal{N}\left(\hat{\mathbf{x}}_{k|k,j-1}^{H_{i}},\mathbf{P}_{k|k,j-1}^{H_{i}}\right)$ , then the MMSE estimator for robot i processing its own analog measurement,  $z_{k}^{i}$ , is identical to the KF.

For the MMSE estimator processing the quantized measurement  $b_k^j$  from robot  $j, j \neq i$ , the state/covariance update equations are given by

$$\hat{\mathbf{x}}_{k|k,j}^{H_i} = \hat{\mathbf{x}}_{k|k,j-1}^{H_i} + \alpha_{H_i}(n) \frac{\mathbf{P}_{k|k,j-1}^{H_i} \mathbf{h}_k^j}{\sqrt{\mathbf{h}_k^{j^T} \mathbf{P}_{k|k,j-1}^{H_i} \mathbf{h}_k^j + \sigma_k^{j^2}}}$$
(5)

$$\mathbf{P}_{k|k,j}^{H_{i}} = \mathbf{P}_{k|k,j-1}^{H_{i}} - \beta_{H_{i}}(n) \frac{\mathbf{P}_{k|k,j-1}^{H_{i}} \mathbf{h}_{k}^{j} \mathbf{h}_{k}^{j}^{T} \mathbf{P}_{k|k,j-1}^{H_{i}}}{\mathbf{h}_{k}^{j}^{T} \mathbf{P}_{k|k,j-1}^{H_{i}} \mathbf{h}_{k}^{j} + \sigma_{k}^{j^{2}}}$$
(6)

where

$$\begin{split} &\alpha_{H_i}(n) = \frac{1}{\sqrt{2\pi}} \frac{\exp\left[-\Delta_{H_i}^2(n)/2\right] - \exp\left[-\Delta_{H_i}^2(n+1)/2\right]}{\mathbb{Q}\left[\Delta_{H_i}(n)\right] - \mathbb{Q}\left[\Delta_{H_i}(n+1)\right]} \\ &\beta_{H_i}(n) = \alpha_{H_i}^2(n) - \frac{1}{\sqrt{2\pi}} \\ &\times \frac{\Delta_{H_i}(n) \exp\left[-\Delta_{H_i}^2(n)/2\right] - \Delta_{H_i}(n+1) \exp\left[-\Delta_{H_i}^2(n+1)/2\right]}{\mathbb{Q}\left[\Delta_{H_i}(n)\right] - \mathbb{Q}\left[\Delta_{H_i}(n+1)\right]} \\ &\Delta_{H_i}(n) = \frac{(\tau_k^j(n) - \mathbf{h}_k^{j^T}(\hat{\mathbf{x}}_{k|k,j-1}^{H_i} - \hat{\mathbf{x}}_{k|k,j-1}^{Q}))}{\sigma_{H_i}} \\ &\sigma_{H_i}^2 = \mathbf{h}_k^{j^T} \mathbf{P}_{k|k,j-1}^{H_i} \mathbf{h}_k^j + \sigma_k^{j^2} \quad \text{and} \quad 0 < \beta_{H_i}(n) < 1. \\ &Proof: \quad \text{See} \ [20] \end{split}$$

3) Quantization thresholds: We now discuss the selection of optimal quantization thresholds,  $\tau_k^j(n)$ ,  $n=2,\ldots,2^f$ , for the batch quantized hybrid estimation framework. For the team to optimally and correctly process quantized measurements received from robot j, robot j's quantization thresholds should be known to the team. Therefore, we choose robot j's thresholds so as to maximize the average reduction in the covariance of the Q-estimator, BQKF. Since the Q-estimator is identical for all robots, every robot in

 $<sup>^5\</sup>mathrm{We}$  assume a round-robin scheduling algorithm where the quantized measurements are generated and processed sequentially, based on robot ids. Therefore, robot i generates and communicates its quantized measurement,  $b_k^i$ , before robot (i+1). Moreover, all robots in the team process  $b_k^i$  to obtain  $\hat{\mathbf{x}}_{k|k,i}^Q$ , before robot (i+1) generates its quantized measurement.

the team can locally calculate the quantization thresholds used by robot j. This optimization problem is equivalent to quantizing the measurement innovation,  $z_k^j - \mathbf{h}_k^{j^T} \hat{\mathbf{x}}_{k|k,j-1}^Q$ , with minimum MSE distortion [16] and its solution is given by the well-known Lloyd-Max quantizer (corresponding values for the optimal quantization thresholds can be found in [21], [22]). Before proceeding, we make the following important observations about the proposed H-BQKF:

- 1) As seen from Proposition 1, even though robot i cannot communicate its analog measurement,  $z_k^i$ , to the team, the H-BQKF enables it to *optimally* process this measurement locally using the KF.
- 2) Define  $\bar{\alpha}_{H_i}(n) = \alpha_{H_i}(n) \sqrt{\mathbf{h}_k^{j^T} \mathbf{P}_{k|k,j-1}^{H_i} \mathbf{h}_k^j + \sigma_k^{j^2}}$ . Using this in (5), we see that the structure of the state update equation, for processing quantized measurements in H-BQKF, is very similar to that of the KF [15]. Moreover, as expected, the measurement innovation in the state update equation of the KF is approximated by  $\bar{\alpha}_{H_i}(n)$  in the H-BQKF.
- 3) The structure of the covariance update equation (see (6)) for processing quantized measurements in the H-BQKF is identical to that of KF, except for the factor  $\beta_{H_i}(n)$ . Since  $0 < \beta_{H_i}(n) < 1$ , the covariance reduction for these estimators will always be less than that of the KF as measurement information is discarded during quantization.
- 4) While processing quantized measurements, the state/covariance update equations for the H-BQKF (see (5) and (6)) are a function of the difference between the predicted measurements,  $\mathbf{h}_k^{J^T} \hat{\mathbf{x}}_{k|k,j-1}^{H_i}$  and  $\mathbf{h}_k^{J^T} \hat{\mathbf{x}}_{k|k,j-1}^{Q}$ , of the H-BQKF and BQKF respectively. Moreover, the covariance reduction in (6) increases as the absolute value of this difference decreases. This is because the quantized measurements encode information about BQKF's measurement innovation. If the difference between the predicted measurements of the H-BQKF and BQKF is large, the quantized measurement will convey very little information to the H-BQKF.
- 5) By choosing f=1 and substituting the corresponding optimal thresholds  $\tau_k^j(1)=-\infty,\ \tau_k^j(2)=0$ , and  $\tau_k^j(3)=\infty$  in Proposition 1, we obtain the special case of the single bit H-estimator in [18].

## B. Iterative Quantization

When additional communication bandwidth becomes available to the robots on-the-fly, robot j can now communicate extra bits, *one bit at a time*, for the same analog measurement  $z_k^j$ . If robot j has communicated (p-1) bits,  $\mathbf{b}_k^{j(1:p-1)}, \ p \geq 1$ , for the analog measurement  $z_k^j$ , then it generates the p-th bit,  $b_k^{j(p)}$ , using the following quantization rule

$$b_k^{j(p)} := \text{sign}[z_k^j - E[z_k^j | \mathbf{b}_{0:k-1}, \mathbf{b}_k^m, \mathbf{b}_k^{j(1:p-1)}]] \quad (7)$$

where,  $\mathbf{b}_{0:k-1}$  denotes the quantized bits from all robots up to time-step k-1, and  $\mathbf{b}_k^m$  denotes the quantized measure-

ments from robot  $m, m = 1, \dots, (j-1)$ , at time-step k. The expected measurement is given by

$$\begin{split} &E[z_{k}^{j}|\mathbf{b}_{0:k-1},\mathbf{b}_{k}^{m},\mathbf{b}_{k}^{j(1:p-1)}]\\ &=\mathbf{h}_{k}^{j^{T}}\hat{\mathbf{x}}_{k|k,j}^{Q(p-1)}+E[v_{k}^{j}|\mathbf{b}_{0:k-1},\mathbf{b}_{k}^{m},\mathbf{b}_{k}^{j(1:p-1)}]. \end{split} \tag{8}$$

Importantly, in the above equation,  $E[v_k^j|\mathbf{b}_{0:k-1},\mathbf{b}_k^m,\mathbf{b}_k^{j(1:p-1)}] \neq 0$ , unless p=1. If p>1, the measurement noise,  $v_k^j$ , is no longer independent of the previous bits,  $\mathbf{b}_k^{j(1:p-1)}$ , since they were generated using the noisy analog measurement,  $z_k^j$ , itself. Therefore, the MMSE estimates of the noise term are needed to correctly generate/decode the bits. These estimates can be obtained by augmenting the state,  $\mathbf{x}_k$ , with the noise term,  $v_k^j$ , and considering the augmented state vector  $\mathbf{x}_k = [\mathbf{x}_k^T, v_k^j]^T$ and a modified  $\mathbf{h}_k^j = [\mathbf{h}_k^j, 1]^T$ . Thus, every time robot i generates its own iteratively quantized measurements or processes iteratively quantized bits communicated by other robots, it has to augment its own state vector with the corresponding measurement noise so that the noise statistics can be correctly estimated. Then the quantization process in (7) becomes identical to that of the sign-of-innovation quantization rule from [14]. Lastly, we note that the Q- and H-estimators for iterative quantization have structure similar to the BQKF and H-BQKF respectively, when the number of quantization bits f = 1. The detailed mathematical expressions are available in [20].

#### IV. SIMULATIONS AND EXPERIMENT

#### A. Simulation Results

The simulation set-up consists of a team of two robots navigating in 2D while performing MC-CL. The continuous-time dynamics for each robot are given by the constant velocity motion model [19]:

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}) + \mathbf{G}_c \begin{bmatrix} w_v \\ w_\omega \end{bmatrix}$$
 (9)

where  $\mathbf{x} = \begin{bmatrix} x, y, \phi, v, \omega \end{bmatrix}^T$ ,  $\mathbf{f}(\mathbf{x}) = \begin{bmatrix} v\cos\phi, v\sin\phi, \omega, 0, 0 \end{bmatrix}^T$  and  $\mathbf{G}_c = \begin{bmatrix} \mathbf{0}_{2\times 3}, \mathbf{I}_{2\times 2} \end{bmatrix}^T$ . The standard deviation of the continuous-time noise in the linear, v, and rotational,  $\omega$ , velocity is chosen to be  $\sigma_v = 0.6325$  m/s. $\sqrt{\mathrm{Hz}}$  and  $\sigma_\omega = 0.4967$  rad/s. $\sqrt{\mathrm{Hz}}$  respectively. Each robot obtains measurements for its linear,  $v_m$ , and rotational,  $\omega_m$ , velocity, as well as its distance,  $d_m$ , and bearing,  $\theta_m$ , to the other robot. The noise in these measurements is modeled as zero-mean, white Gaussian with standard deviation  $\sigma_{v_m} = 0.07$  m/s,  $\sigma_{\omega_m} = 0.28$  rad/s for the linear and rotational velocity measurements, respectively, and  $\sigma_{d_m} = 0.05$  m,  $\sigma_{\theta_m} = 0.09$  rad for the corresponding distance and bearing measurements.

In this section, we compare the performance of the proposed H-estimators (local analog and remote quantized), H-BQKF and H-IQKF with: (1) the Q-estimators (local and remote quantized measurements), BQKF and IQKF, using 1-4 bits per analog measurement, and (2) the standard EKF that uses analog measurements from all robots and hence is our benchmark. Figures 2, 3 show the root mean squared

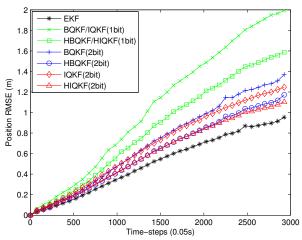


Fig. 2. Comparison of position RMSE for EKF, and 1-2 bit H-BQKF, H-IQKF, BQKF and IQKF.

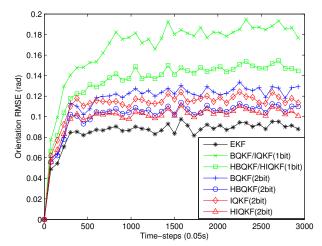


Fig. 3. Comparison of orientation RMSE for EKF, and 1-2 bit H-BQKF, H-IQKF, BQKF and IQKF.

error (RMSE) in the position and orientation estimates for these estimators, averaged over the 2 robots and 100 Monte Carlo trials. For clarity, we have included only the results for n=1,2 bits. Since the estimates generated by the Hestimator are different for each robot, the RMSE for the Hestimators, H-BQKF and H-IQKF, are also averaged over estimators maintained by each robot. Table I, presents the results for position and orientation RMSE, for n=1,2,4 bits, averaged over the duration of the simulation run. Moreover, since the 1-bit iterative- and batch-quantized estimators are identical, Table I omits the results for the 1-bit iteratively quantized filters.

From Figs. 2, 3, and Table I, we observe that the estimates generated by the proposed H-estimators, H-BQKF and H-IQKF, are more accurate that their Q-estimator counterparts, BQKF and IQKF, irrespective of the number of quantization bits (n =  $\{1,2,4\}$ ) considered. Specifically, the 1-bit hybrid filters are 20% more accurate than the 1-bit quantized filters, while the 2-bit hybrid filters show a performance improvement of 13% over their quantized counterparts. Overall, the

error in the estimates decreases as we increase the number of quantization bits and by communicating as few as 4 bits per analog measurement, both the H- and Q-estimators are able to achieve accuracy very close to that of the analog EKF. Also, for a fixed number of bits, the performance of both the batch and iteratively quantized estimators is comparable. Thus, we conclude that by including their local analog measurements in the estimation process, without any additional communication overhead, the robots are able to substantially improve the estimation accuracy of CL.

 $\label{eq:table in table in$ 

|                 | Pos. RMSE (m) | Orient. RMSE (rad) |
|-----------------|---------------|--------------------|
| BQKF (1 bit):   | 1.0743        | 0.1689             |
| H-BQKF (1 bit): | 0.8584        | 0.1361             |
| BQKF (2 bit):   | 0.7203        | 0.1171             |
| H-BQKF (2 bit): | 0.6199        | 0.1018             |
| BQKF (4 bit):   | 0.5515        | 0.0910             |
| H-BQKF (4 bit): | 0.5337        | 0.0885             |
| IQKF (2 bit):   | 0.6932        | 0.1120             |
| H-IQKF (2 bit): | 0.6105        | 0.0997             |
| IQKF (4 bit):   | 0.6176        | 0.1007             |
| H-IQKF (4 bit): | 0.5673        | 0.0934             |
| EKF (analog):   | 0.5151        | 0.0858             |

#### B. Experimental Results

Experimental validation was carried out using a team of four Pioneer-I robots moving in a rectangular arena of 4 m  $\times$  2.5 m for approximately 16 minutes. An overhead camera is used to obtain the robots' poses in a global coordinate frame (ground truth).

The robots move with a constant velocity of 0.1 m/s while avoiding collisions with boundaries of the arena and other robots in the team. The robots obtain linear and rotational velocity (odometry) measurements, and relative distance and bearing measurements at a frequency of 1 Hz. The noise standard deviations of the odometry measurements for the heterogeneous robot team vary from 0.0078 rad/s to 0.02 rad/s for rotational velocity, and from 0.0032 m/s to 0.0059 m/s for linear velocity. The relative distance and bearing measurements between the robots are generated synthetically using data from the overhead camera and adding Gaussian noise with standard deviation  $\sigma_d = 0.05$  m for distance and  $\sigma_{\theta} = 2$  deg for relative bearing. Table II, presents the position and orientation RMSE for each robot, averaged over all time-steps. For the H-estimators, these quantities are also averaged across all robot's H-estimators. The experiment corroborates our simulation results. From the RMSE data in the table, we conclude that the n-bit, n ={1,2,3}, H-estimators (H-BQKF and H-IQKF) outperform the corresponding Q-estimators (BQKF and IQKF). Thus the H-estimators, by enabling robots to include their local analog measurements in the estimation process, significantly improve the estimation accuracy of CL. Specifically, the improvement in performance of the H-estimators over the Qestimators is more pronounced for the n = 1, 2 bit scenario,

TABLE II
EXPERIMENTAL RESULTS FOR THE HYBRID ESTIMATION FRAMEWORK

| RMS Position error (m)      |         |         |         |         |  |
|-----------------------------|---------|---------|---------|---------|--|
|                             | Robot 1 | Robot 2 | Robot 3 | Robot 4 |  |
| BQKF (1 bit):               | 1.3643  | 1.1973  | 1.2849  | 1.1794  |  |
| H-BQKF (1 bit):             | 1.0254  | 0.9358  | 0.9716  | 0.9714  |  |
| BQKF (2 bit):               | 0.8654  | 0.7405  | 0.8383  | 0.8752  |  |
| H-BQKF (2 bit):             | 0.7516  | 0.6652  | 0.7254  | 0.7542  |  |
| BQKF (3 bit):               | 0.5330  | 0.4824  | 0.5168  | 0.4742  |  |
| H-BQKF (3 bit):             | 0.5821  | 0.5553  | 0.5665  | 0.5809  |  |
| IQKF (2 bit):               | 0.5990  | 0.5961  | 0.5758  | 0.7062  |  |
| H-IQKF (2 bit):             | 0.6023  | 0.5879  | 0.5870  | 0.6637  |  |
| IQKF (3 bit):               | 0.6255  | 0.6187  | 0.6291  | 0.7028  |  |
| H-IQKF (3 bit):             | 0.5770  | 0.5547  | 0.5650  | 0.5974  |  |
| EKF (analog):               | 0.5651  | 0.5385  | 0.5239  | 0.5719  |  |
| RMS Orientation error (rad) |         |         |         |         |  |
|                             | Robot 1 | Robot 2 | Robot 3 | Robot 4 |  |
| BQKF (1 bit):               | 0.7793  | 0.7810  | 0.7777  | 0.7820  |  |
| H-BQKF (1 bit):             | 0.6331  | 0.6337  | 0.6289  | 0.6345  |  |
| BQKF (2 bit):               | 0.6305  | 0.6314  | 0.6294  | 0.6296  |  |
| H-BQKF (2 bit):             | 0.5392  | 0.5402  | 0.5386  | 0.5385  |  |
| BQKF (3 bit):               | 0.3182  | 0.3174  | 0.3171  | 0.3176  |  |
| H-BQKF (3 bit):             | 0.4064  | 0.4062  | 0.4059  | 0.4058  |  |
| IQKF (2 bit):               | 0.4530  | 0.4549  | 0.4540  | 0.4543  |  |
| H-IQKF (2 bit):             | 0.4300  | 0.4316  | 0.4306  | 0.4314  |  |
| IQKF (3 bit):               | 0.4480  | 0.4468  | 0.4461  | 0.4465  |  |
| H-IQKF (3 bit):             | 0.3812  | 0.3802  | 0.3795  | 0.3796  |  |
| EKF (analog):               | 0.4055  | 0.4064  | 0.4059  | 0.4060  |  |

while with n=3 bits, the performance of both the Q- and H-estimators is very close to that of the standard analog EKF. For this particular experiment, we observe that in some cases, the RMSE for the 3-bit Q- and H-estimators is slightly lower than the EKF. However, these results are reported for a single experimental run, and in general, we would expect the analog EKF to outperform all the Q- and H-estimators.

#### V. CONCLUSIONS AND FUTURE WORK

In this paper, we derived two MMSE estimators for the hybrid estimation framework, the H-BQKF and the H-IQKF, that can process local analog measurements along with  $f \geq 1$  bits per remote analog measurement, in order to improve the estimation accuracy of CL under time-varying communication-bandwidth constraints. We tested the performance and accuracy of the proposed multi-bit hybrid filters for the CL application in simulations and experiment and showed that they outperform the existing multi-bit quantized filters, BQKF and IQKF. As part of our future work, we plan to analytical evaluate the performance of the proposed H-BQKF and the H-IQKF. Furthermore, we will also rigorously analyze the effect of packet-loss and communication errors on the performance of the proposed estimators.

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